ULTRALOW-NOISE, HIGH-PSRR, FAST, RF, 500-mA LOW-DROPOUT LINEAR REGULATORS

FEATURES

- **500-mA Low-Dropout Regulator With Enable**
- **Available in Fixed and Adjustable (1.2-V to 5.5-V) Versions**
- **High PSRR (50 dB at 10 kHz)**
- **Ultralow Noise (33 μV_{RMS}, TPS79530)**
- **Fast Start-Up Time (50 μs)**
- **Stable With a 1-μF Ceramic Capacitor**
- **Excellent Load/Line Transient Response**
- **Very Low Dropout Voltage (110 mV at Full Load, TPS79530)**
- **6-Pin SOT223 and 3 $\times$ 3 SON Packages**

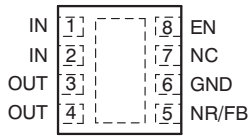
APPLICATIONS

- **RF: VCOs, Receivers, ADCs**
- **Audio**
- **Bluetooth®, Wireless LAN**
- **Cellular and Cordless Telephones**
- **Handheld Organizers, PDAs**

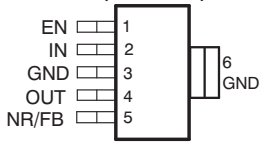
DESCRIPTION

The TPS795xx family of low-dropout (LDO), low-power linear voltage regulators features high power-supply rejection ratio (PSRR), ultralow noise, fast start-up, and excellent line and load transient responses in small outline, SOT223-6 and 3 x 3 SON packages. Each device in the family is stable with a small 1-μF ceramic capacitor on the output. The family uses an advanced, proprietary BiCMOS fabrication process to yield extremely low dropout voltages (for example, 110 mV at 500 mA). Each device achieves fast start-up times (approximately 50 μs with a 0.001-μF bypass capacitor) while consuming very low quiescent current (265 μA, typical). Moreover, when the device is placed in standby mode, the supply current is reduced to less than 1 μA. The TPS79530 exhibits approximately 33 μV_{RMS} of output voltage noise at 3.0 V output with a 0.1-μF bypass capacitor. Applications with analog components that are noise-sensitive, such as portable RF electronics, benefit from the high-PSRR and low-noise features, as well as from the fast response time.

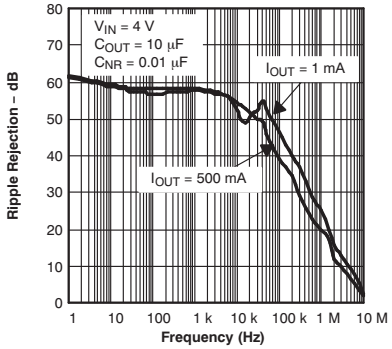
DRB PACKAGE
3mm x 3mm SON
(TOP VIEW)



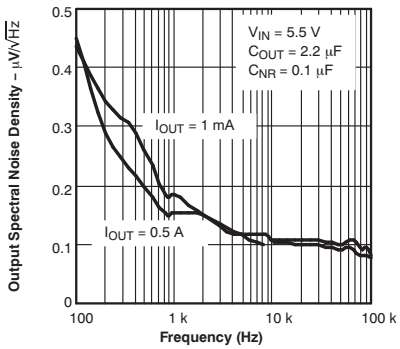
DCQ PACKAGE
SOT223-6
(TOP VIEW)



TPS79530
RIPPLE REJECTION vs FREQUENCY



TPS79530
OUTPUT SPECTRAL NOISE DENSITY vs FREQUENCY



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

ORDERING INFORMATION(1)

PRODUCT	V_{OUT}(2)
TPS795**xxyyyz**	**XX** is nominal output voltage (for example, 28 = 2.8 V, 285 = 2.85 V, 01 = Adjustable). **YYY** is package designator. **Z** is package quantity.

(1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI website at www.ti.com.
(2) Output voltages from 1.3 V to 5.0 V in 100 mV increments are available; minimum order quantities may apply. Contact factory for details and availability.

ABSOLUTE MAXIMUM RATINGS

over operating temperature (unless otherwise noted)(1)

	VALUE
V_{IN} range	– 0.3 V to 6 V
V_{EN} range	–0.3 V to V_{IN} + 0.3 V
V_{OUT} range	6 V
Peak output current	Internally limited
ESD rating, HBM	2 kV
ESD rating, CDM	500 V
Continuous total power dissipation	See Dissipation Rating Table
Junction temperature range, T_J	–40°C to +150°C
Storage temperature range, T_{stg}	–65°C to +150°C

(1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not implied.

DISSIPATION RATING TABLE

PACKAGE	BOARD	$R_{\theta JC}$	$R_{\theta JA}$
SOT223	Low K(1)	15°C/W	53°C/W
3 x 3 SON	High-K(2)	1.2°C/W	40°C/W

(1) The JEDEC low-K (1s) board design used to derive this data was a 3-inch × 3-inch (7.5 cm × 7.5cm), two-layer board with 2-ounce copper traces on top of the board.
(2) The JEDEC high-K (2s2p) board design used to derive this data was a 3-inch × 3-inch (7,5-cm × 7,5-cm), multilayer board with 1-ounce internal power and ground planes and 2-ounce copper traces on top and bottom of the board.

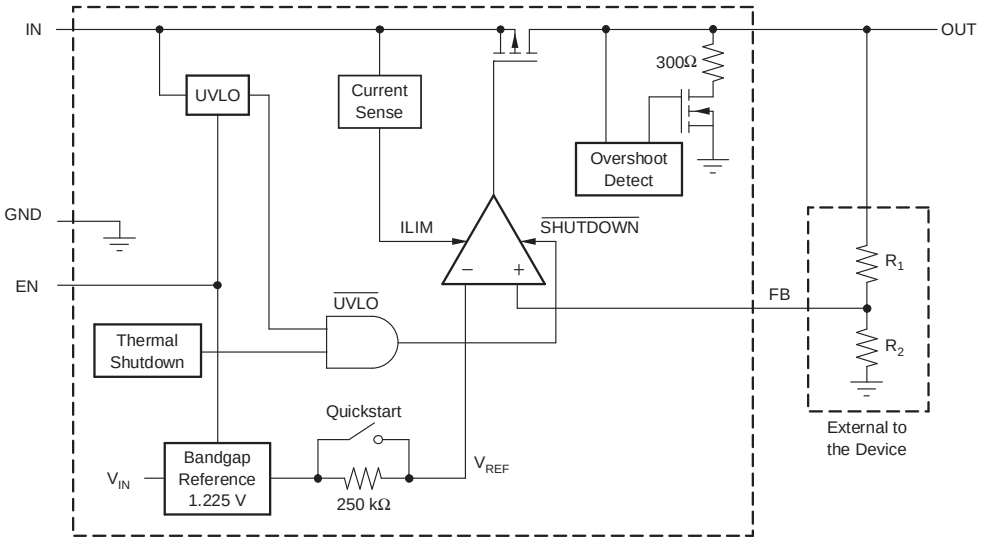
ELECTRICAL CHARACTERISTICS

Over recommended operating temperature range (T_J = –40°C to +125°C), $V_{EN} = V_{IN}$, $V_{IN} = V_{OUT(nom)}$ + 1 V(1), I_{OUT} = 1 mA, C_{OUT} = 10 µF, C_{NR} = 0.01 µF, unless otherwise noted. Typical values are at +25°C.

PARAMETER			TEST CONDITIONS		MIN	TYP	MAX	UNIT
Input voltage, V_{IN}(1)					2.7		5.5	V
Internal reference, V_{FB} (TPS79501)					1.200	1.225	1.250	V
Continuous output current, I_{OUT}					0		500	mA
Output voltage	Output voltage range	TPS79501			1.225		$5.5 – V_{DO}$	V
	Accuracy	TPS79501(2)	0 µA ≤ I_{OUT} ≤ 500 mA, V_{OUT} + 1 V ≤ V_{IN} ≤ 5.5 V(1)		0.98(V_{OUT})	V_{OUT}	1.02(V_{OUT})	V
		Fixed V_{OUT}	0 µA ≤ I_{OUT} ≤ 500 mA, V_{OUT} + 1 V ≤ V_{IN} ≤ 5.5 V(1)		–2.0		+2.0	%
Output voltage line regulation ($\Delta V_{OUT}\%/\Delta V_{IN}$)(1)			V_{OUT} + 1 V ≤ V_{IN} ≤ 5.5 V			0.05	0.12	%/V
Load regulation ($\Delta V_{OUT}\%/\Delta I_{OUT}$)			0 µA ≤ I_{OUT} ≤ 500 mA,			3		mV
Dropout voltage(3) $V_{IN} = V_{OUT(nom)}$ - 0.1 V		TPS79530	I_{OUT} = 500 mA			110	170	mV
		TPS79533	I_{OUT} = 500 mA			105	160	
Output current limit			V_{OUT} = 0 V		2.4	2.8	4.2	A
Ground pin current			0 µA ≤ I_{OUT} ≤ 500 mA			265	385	µA
Shutdown current(4)			V_{EN} = 0 V, 2.7 V ≤ V_{IN} ≤ 5.5 V			0.07	1	µA
FB pin current			V_{FB} = 1.225 V				1	µA
Power-supply ripple rejection		TPS79530	f = 100 Hz, I_{OUT} = 10 mA			59		dB
			f = 100 Hz, I_{OUT} = 500 mA			58		
			f = 10 kHz, I_{OUT} = 500 mA			50		
			f = 100 kHz, I_{OUT} = 500 mA			39		
Output noise voltage (TPS79530)			BW = 100 Hz to 100 kHz, I_{OUT} = 500 mA	C_{NR} = 0.001 µF		46		μV_{RMS}
				C_{NR} = 0.0047 µF		41		
				C_{NR} = 0.01 µF		35		
				C_{NR} = 0.1 µF		33		
Time, start-up (TPS79530)			R_L = 6 Ω, C_{OUT} = 1 µF	C_{NR} = 0.001 µF		50		µs
				C_{NR} = 0.0047 µF		75		
				C_{NR} = 0.01 µF		110		
High-level enable input voltage			2.7 V ≤ V_{IN} ≤ 5.5 V		1.7		V_{IN}	V
Low-level enable input voltage			2.7 V ≤ V_{IN} ≤ 5.5 V				0.7	V
EN pin current			V_{EN} = 0 V		1		1	µA
UVLO threshold			V_{CC} rising		2.25		2.65	V
UVLO hysteresis						100		mV

(1) Minimum V_{IN} is 2.7 V or V_{OUT} + V_{DO}, whichever is greater.
(2) Tolerance of external resistors not included in this specification.
(3) Dropout is not measured for the TPS79501 and TPS79525 since minimum V_{IN} = 2.7 V.
(4) For adjustable version, this applies only after V_{IN} is applied; then V_{EN} transitions high to low.

FUNCTIONAL BLOCK DIAGRAM—ADJUSTABLE VERSION



FUNCTIONAL BLOCK DIAGRAM—FIXED VERSION

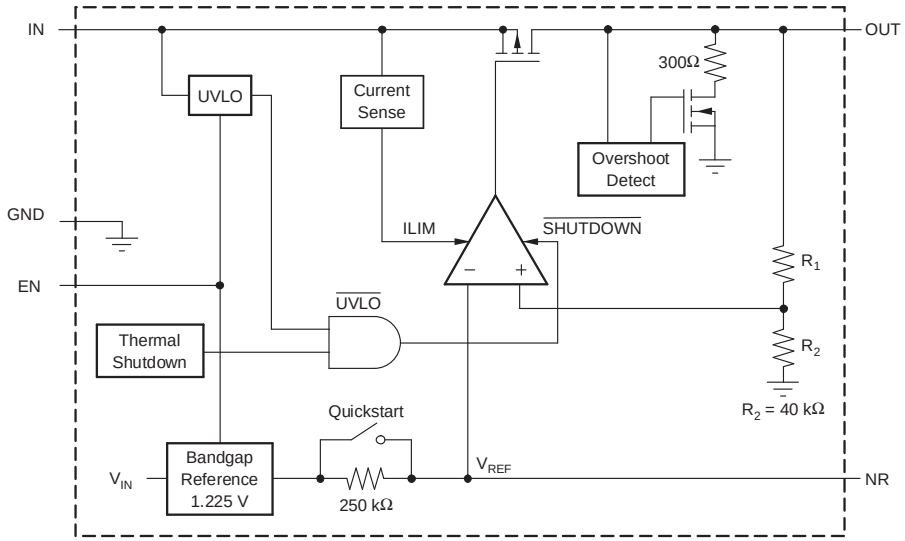


Table 1. Terminal Functions

NAME	SOT223 (DCQ) PIN NO.	3x3 SON (DRB) PIN NO.	DESCRIPTION
IN	2	1, 2	Unregulated input to the device
GND	3, 6	6	Regulator ground
EN	1	8	Driving the enable pin (EN) high turns on the regulator. Driving this pin low puts the regulator into shutdown mode. EN can be connected to IN if not used.
NR	5	5	Noise-reduction pin for fixed versions only. Connecting an external capacitor to this pin bypasses noise generated by the internal bandgap, which improves power-supply rejection and reduces output noise. (Not available on adjustable versions.)
FB	5	5	Feedback input voltage for the adjustable device. (Not available on fixed voltage versions.)
OUT	4	3, 4	Regulator output.
NC	–	7	Not connected

TYPICAL CHARACTERISTICS

TPS79530
OUTPUT VOLTAGE
vs OUTPUT CURRENT

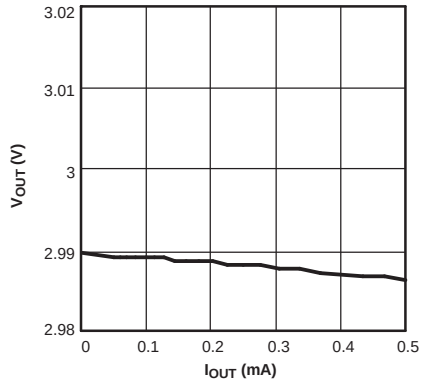


Figure 1.

TPS79530
OUTPUT VOLTAGE
vs JUNCTION TEMPERATURE

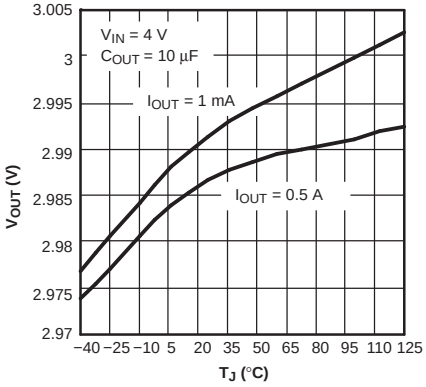


Figure 2.

TPS79530
GROUND CURRENT
vs JUNCTION TEMPERATURE

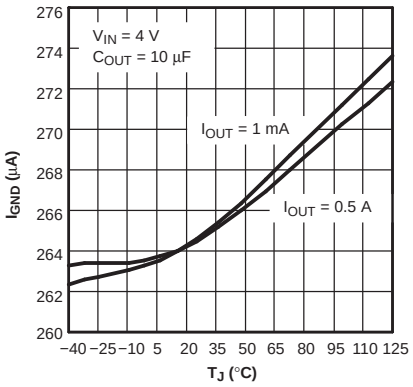


Figure 3.

TPS79530
OUTPUT SPECTRAL
NOISE DENSITY
vs FREQUENCY

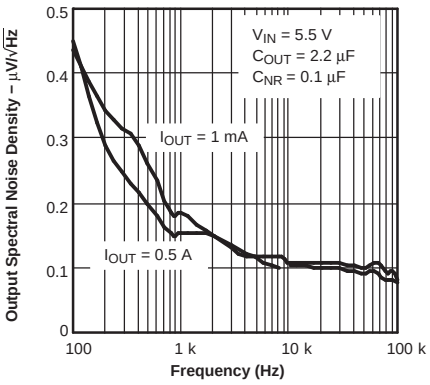


Figure 4.

TPS79530
OUTPUT SPECTRAL
NOISE DENSITY
vs FREQUENCY

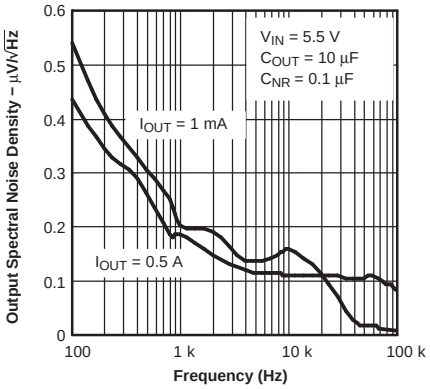


Figure 5.

TPS79530
OUTPUT SPECTRAL
NOISE DENSITY
vs FREQUENCY

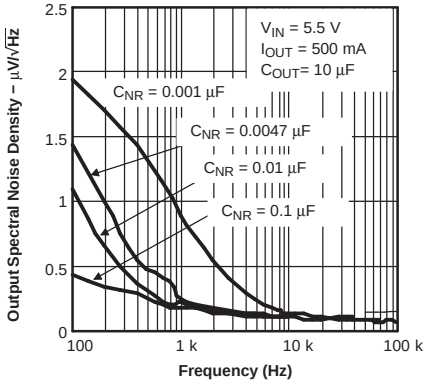


Figure 6.

TPS79530
ROOT MEAN SQUARED
OUTPUT NOISE vs C_{NR}

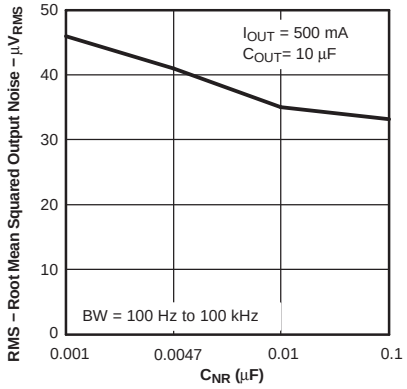


Figure 7.

TPS79530
DROPOUT VOLTAGE
vs JUNCTION TEMPERATURE

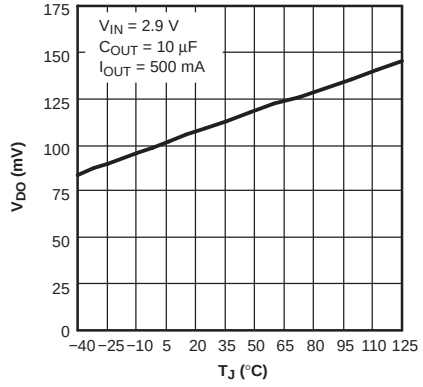


Figure 8.

TPS79530
RIPPLE REJECTION
vs FREQUENCY

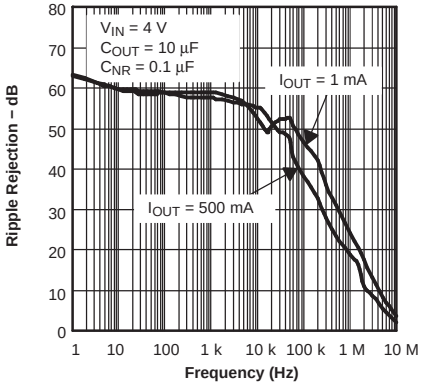


Figure 9.

TYPICAL CHARACTERISTICS (continued)

TPS79530
RIPPLE REJECTION
vs FREQUENCY

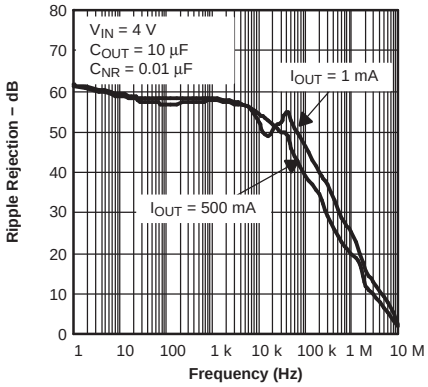


Figure 10.

TPS79530
RIPPLE REJECTION
vs FREQUENCY

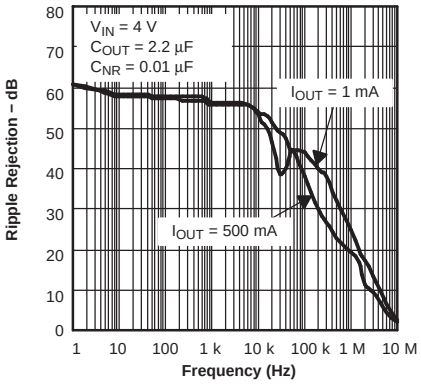


Figure 11.

TPS79530
RIPPLE REJECTION
vs FREQUENCY

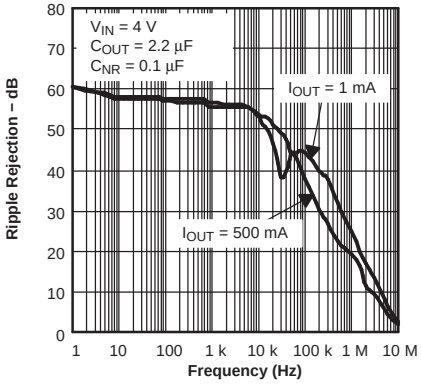


Figure 12.

TPS79530
START-UP TIME

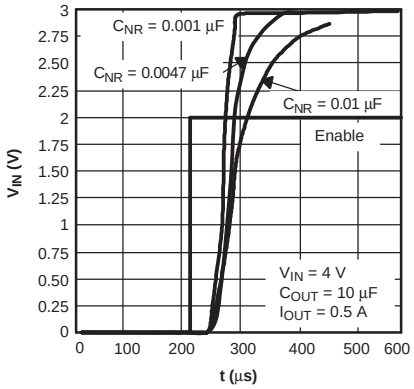


Figure 13.

TPS79518
LINE TRANSIENT RESPONSE

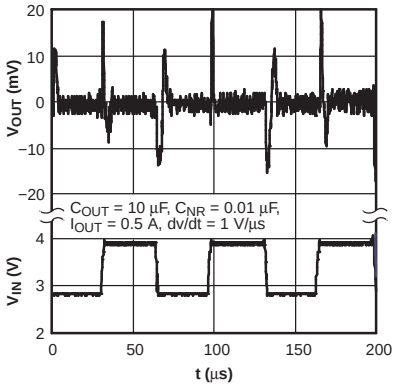


Figure 14.

TPS79530
LINE TRANSIENT RESPONSE

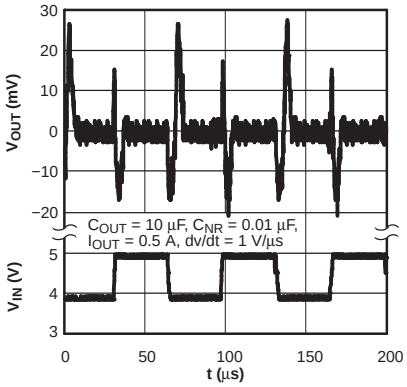


Figure 15.

TPS79530
LOAD TRANSIENT RESPONSE

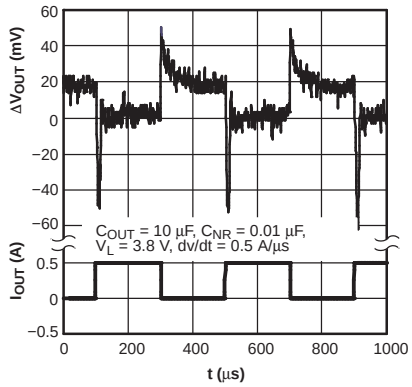


Figure 16.

TPS79525
POWER UP/POWER DOWN

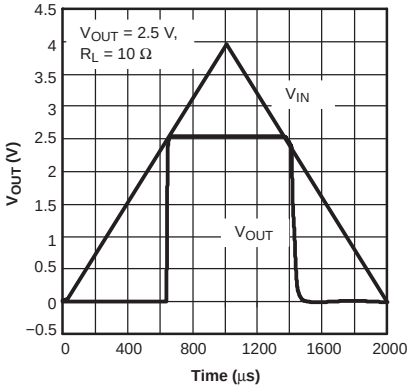


Figure 17.

TPS79530
DROPOUT VOLTAGE
vs OUTPUT CURRENT

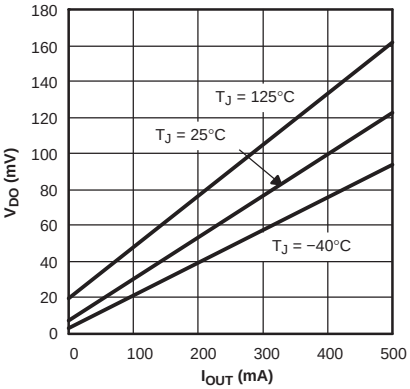


Figure 18.

TYPICAL CHARACTERISTICS (continued)

TPS79501
DROPOUT VOLTAGE
vs INPUT VOLTAGE

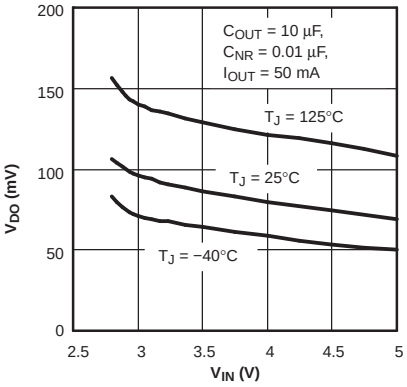


Figure 19.

TPS79530
TYPICAL REGIONS OF STABILITY
EQUIVALENT SERIES RESISTANCE (ESR)
vs OUTPUT CURRENT

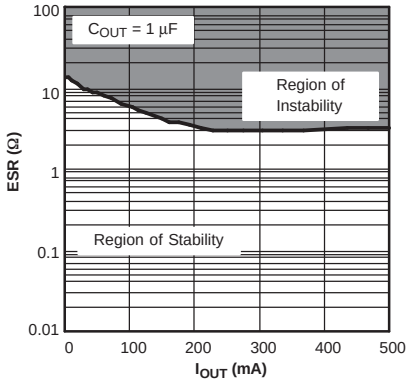


Figure 20.

TPS79530
TYPICAL REGIONS OF STABILITY
EQUIVALENT SERIES RESISTANCE (ESR)
vs OUTPUT CURRENT

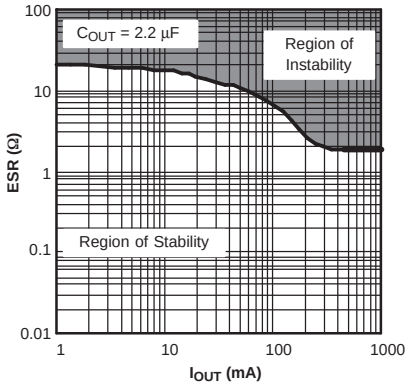


Figure 21.

TPS79530
TYPICAL REGIONS OF STABILITY
EQUIVALENT SERIES RESISTANCE (ESR)
vs OUTPUT CURRENT

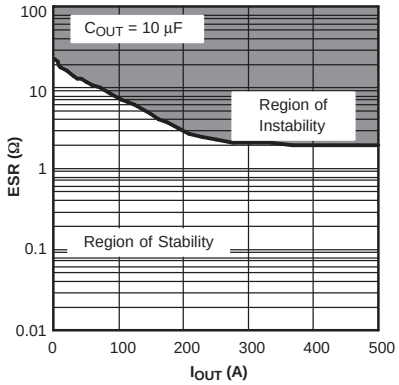


Figure 22.

APPLICATION INFORMATION

The TPS795xx family of low-dropout (LDO) regulators has been optimized for use in noise-sensitive equipment. The device features extremely low dropout voltages, high PSRR, ultralow output noise, low quiescent current (265 μA typically), and an enable input to reduce supply currents to less than 1 μA when the regulator is turned off.

A typical application circuit is shown in Figure 23.

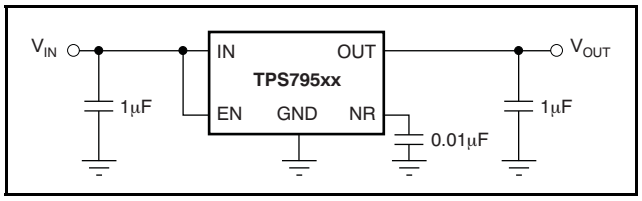


Figure 23. Typical Application Circuit

EXTERNAL CAPACITOR REQUIREMENTS

Although not required, it is good analog design practice to place a 0.1μF — 2.2μF capacitor near the input of the regulator to counteract reactive input sources. A higher-value input capacitor may be necessary if large, fast-rise-time load transients are anticipated and the device is located several inches from the power source.

Like most low-dropout regulators, the TPS795xx requires an output capacitor connected between OUT and GND to stabilize the internal control loop. The minimum recommended capacitor is 1 μF. Any 1 μF or larger ceramic capacitor is suitable.

The internal voltage reference is a key source of noise in an LDO regulator. The TPS795xx has an NR pin which is connected to the voltage reference through a 250-kΩ internal resistor. The 250-kΩ internal resistor, in conjunction with an external bypass capacitor connected to the NR pin, creates a low-pass filter to reduce the voltage reference noise and, therefore, the noise at the regulator output. In order for the regulator to operate properly, the current flow out of the NR pin must be at a minimum, because any leakage current creates an IR drop across the internal resistor, thus creating an output error. Therefore, the bypass capacitor must have minimal leakage current. The bypass capacitor should be no more than 0.1-μF in order to ensure that it is fully charged during the quickstart time provided by the internal switch shown in the Functional Block Diagram.

For example, the TPS79530 exhibits only 33 μV_{RMS} of output voltage noise using a 0.1-μF ceramic bypass capacitor and a 10-μF ceramic output capacitor. Note that the output starts up slower as the bypass capacitance increases because of the RC time constant at the bypass pin that is created by the internal 250-kΩ resistor and external capacitor.

BOARD LAYOUT RECOMMENDATION TO IMPROVE PSRR AND NOISE PERFORMANCE

To improve ac measurements such as PSRR, output noise, and transient response, it is recommended that the board be designed with separate ground planes for V_{IN} and V_{OUT}, with each ground plane connected only at the ground pin of the device. In addition, the ground connection for the bypass capacitor should connect directly to the ground pin of the device.

REGULATOR MOUNTING

The tab of the SOT223-6 package is electrically connected to ground. For best thermal performance, the tab of the surface-mount version should be soldered directly to a circuit-board copper area. Increasing the copper area improves heat dissipation.

Solder pad footprint recommendations for the devices are presented in Application Report SBFA015, *Solder Pad Recommendations for Surface-Mount Devices*, available from the TI web site (www.ti.com).

PROGRAMMING THE TPS79501 ADJUSTABLE LDO REGULATOR

The output voltage of the TPS79501 adjustable regulator is programmed using an external resistor divider as shown in Figure 24. The output voltage is calculated using Equation 1:

$$V_{OUT} = V_{REF} \times \left(1 + \frac{R_1}{R_2}\right) \quad (1)$$

where:

- V_{REF} = 1.2246 V typ (the internal reference voltage)

Resistors R_1 and R_2 should be chosen for approximately 40-μA divider current. Lower value resistors can be used for improved noise performance, but the device wastes more power. Higher values should be avoided, as leakage current at FB increases the output voltage error.

The recommended design procedure is to choose R_2 = 30.1 kΩ to set the divider current at 40 μA, C_1 = 15 pF for stability, and then calculate R_1 using Equation 2:

$$R_1 = \left(\frac{V_{OUT}}{V_{REF}} - 1\right) \times R_2 \quad (2)$$

In order to improve the stability of the adjustable version, it is suggested that a small compensation capacitor be placed between OUT and FB.

The approximate value of this capacitor can be calculated as Equation 3:

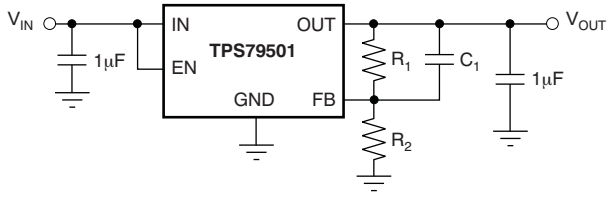
$$C_1 = \frac{(3 \times 10^{-7}) \times (R_1 + R_2)}{(R_1 \times R_2)} \quad (3)$$

The suggested value of this capacitor for several resistor ratios is shown in the table within Figure 24. If this capacitor is not used (such as in a unity-gain configuration), then the minimum recommended output capacitor is 2.2 μF instead of 1 μF.

REGULATOR PROTECTION

The TPS795xx PMOS-pass transistor has a built-in back diode that conducts reverse current when the input voltage drops below the output voltage (for example, during power down). Current is conducted from the output to the input and is not internally limited. If extended reverse voltage operation is anticipated, external limiting might be appropriate.

The TPS795xx features internal current limiting and thermal protection. During normal operation, the TPS795xx limits output current to approximately 2.8 A. When current limiting engages, the output voltage scales back linearly until the overcurrent condition ends. While current limiting is designed to prevent gross device failure, care should be taken not to exceed the power dissipation ratings of the package. If the temperature of the device exceeds approximately 165°C, thermal-protection circuitry shuts it down. Once the device has cooled down to below approximately 140°C, regulator operation resumes.



OUTPUT VOLTAGE PROGRAMMING GUIDE

OUTPUT VOLTAGE	R_1	R_2	C_1
1.8 V	14.0 kΩ	30.1 kΩ	33 pF
3.6 V	57.9 kΩ	30.1 kΩ	15 pF

Figure 24. TPS79501 Adjustable LDO Regulator Programming

THERMAL INFORMATION

The amount of heat that an LDO linear regulator generates is directly proportional to the amount of power it dissipates during operation. All integrated circuits have a maximum allowable junction temperature ($T_J$max) above which normal operation is not assured. A system designer must design the operating environment so that the operating junction temperature (T_J) does not exceed the maximum junction temperature ($T_J$max). The two main environmental variables that a designer can use to improve thermal performance are air flow and external heatsinks. The purpose of this information is to aid the designer in determining the proper operating environment for a linear regulator that is operating at a specific power level.

In general, the maximum expected power ($P_D$max) consumed by a linear regulator is computed as shown in Equation 4:

$$P_D\text{max} = \left(V_{IN(avg)} - V_{OUT(avg)}\right) \times I_{OUT(avg)} + V_{I(avg)} \times I_Q \quad (4)$$

where:

- $V_{IN(avg)}$ is the average input voltage
- $V_{OUT(avg)}$ is the average output voltage
- $I_{OUT(avg)}$ is the average output current
- I_Q is the quiescent current

For most TI LDO regulators, the quiescent current is insignificant compared to the average output current; therefore, the term $V_{IN(avg)}$ x I_Q can be neglected. The operating junction temperature is computed by adding the ambient temperature (T_A) and the increase in temperature due to the regulator's power dissipation. The temperature rise is computed by multiplying the maximum expected power dissipation by the sum of the thermal resistances between the junction and the case ($R_{\Theta JC}$), the case to heatsink ($R_{\Theta CS}$), and the heatsink to ambient ($R_{\Theta SA}$). Thermal resistances are measures of how effectively an object dissipates heat. Typically, the larger the device, the more surface area available for power dissipation and the lower the object's thermal resistance.

Figure 25 illustrates these thermal resistances for a SOT223 package mounted in a JEDEC low-K board.

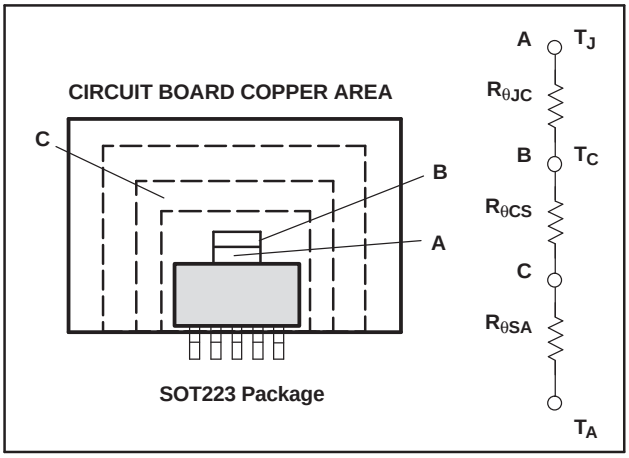


Figure 25. Thermal Resistances

Equation 5 summarizes the computation:

$$T_J = T_A + P_D\text{max} \times \left(R_{\theta JC} + R_{\theta CS} + R_{\theta SA}\right) \quad (5)$$

The $R_{\Theta JC}$ is specific to each regulator as determined by its package, lead frame, and die size provided in the regulator's data sheet. The $R_{\Theta SA}$ is a function of the type and size of heatsink. For example, *black body radiator* type heatsinks can have $R_{\Theta CS}$ values ranging from 5°C/W for very large heatsinks to 50°C/W for very small heatsinks. The $R_{\Theta CS}$ is a function of how the package is attached to the heatsink. For example, if a thermal compound is used to attach a heatsink to a SOT223 package, $R_{\Theta CS}$ of 1°C/W is reasonable.

Even if no external *black body radiator* type heatsink is attached to the package, the board on which the regulator is mounted provides some heatsinking through the pin solder connections. Some packages, like the DDPAK and SOT223 packages, use a copper plane underneath the package or the circuit board ground plane for additional heatsinking to improve their thermal performance. Computer-aided thermal modeling can be used to compute very accurate approximations of an integrated circuit's thermal performance in different operating environments (for example, different types of circuit boards, different types and sizes of heatsinks, different air flows, etc.). Using these models, the three thermal resistances can be combined into one thermal resistance between junction and ambient ($R_{\Theta JA}$). This $R_{\Theta JA}$ is valid only for the specific operating environment used in the computer model.

Equation 5 simplifies into Equation 6:

$$T_J = T_A + P_D max \times R_{\theta JA} \quad (6)$$

Rearranging Equation 6 gives Equation 7:

$$R_{\theta JA} = \frac{T_J - T_A}{P_D max} \quad (7)$$

Using Equation 6 and the computer model generated curves shown in Figure 26, a designer can quickly compute the required heatsink thermal resistance/board area for a given ambient temperature, power dissipation, and operating environment.

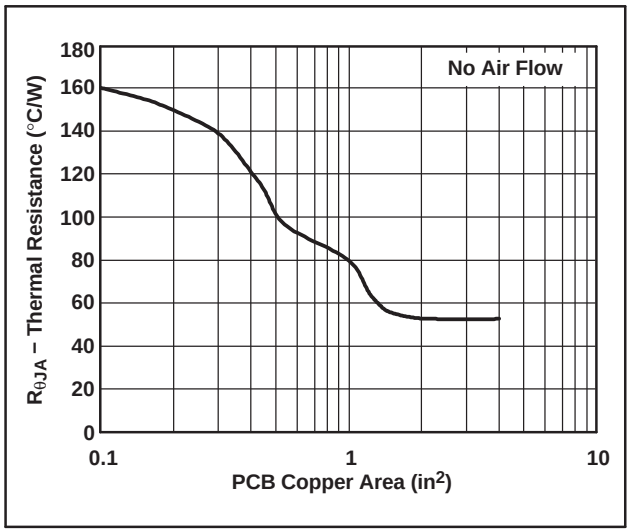


Figure 26. SOT223 Thermal Resistance vs PCB Copper Area

SOT223 POWER DISSIPATION

The SOT223 package provides an effective means of managing power dissipation in surface-mount applications. The SOT223 package dimensions are provided in the *Mechanical Data* section at the end of the data sheet. The addition of a copper plane directly underneath the SOT223 package enhances the thermal performance of the package.

To illustrate, the TPS79525 in a SOT223 package was chosen. For this example, the average input voltage is 3.3 V, the output voltage is 2.5 V, the average output current is 1 A, the ambient temperature 55°C, no air flow is present, and the operating environment is the same as documented below. Neglecting the quiescent current, the maximum average power is Equation 8:

$$P_D max = (3.3 - 2.5)V \times 1A = 800mW \quad (8)$$

Substituting $T_J max$ for T_J into Equation 4 gives Equation 9:

$$R_{\theta JA} max = (125 - 55)°C/800mW = 87.5°C/W \quad (9)$$

From Figure 26, $R_{\theta JA}$ vs PCB Copper Area, the ground plane needs to be 0.55 in^2 for the part to dissipate 800 mW. The operating environment used to construct Figure 26 consisted of a board with 1 oz. copper planes. The package is soldered to a 1 oz. copper pad on the top of the board. The pad is tied through thermal vias to the 1 oz. ground plane.

From the data in Figure 26 and rearranging equation 6, the maximum power dissipation for a different ground plane area and a specific ambient temperature can be computed, as shown in Figure 27.

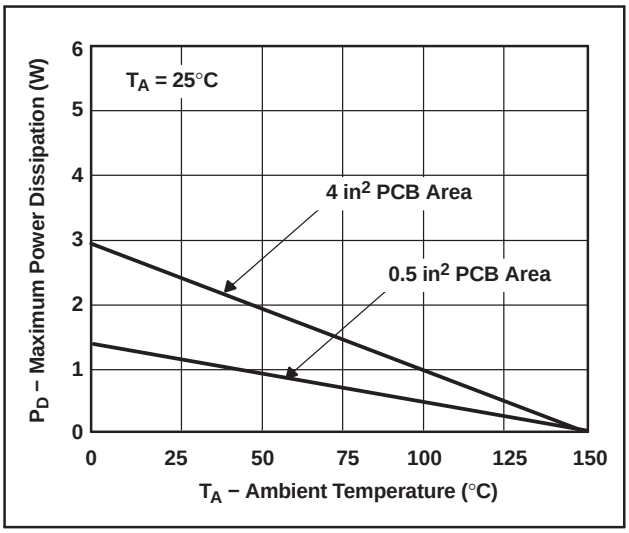


Figure 27. SOT223 Maximum Power Dissipation vs Ambient Temperature

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)
TPS79501DCQ	ACTIVE	SOT-223	DCQ	6	78	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79501DCQG4	ACTIVE	SOT-223	DCQ	6	78	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79501DCQR	ACTIVE	SOT-223	DCQ	6	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79501DCQRG4	ACTIVE	SOT-223	DCQ	6	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79501DRBR	ACTIVE	SON	DRB	8	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79501DRBRG4	ACTIVE	SON	DRB	8	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79501DRBT	ACTIVE	SON	DRB	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79501DRBTG4	ACTIVE	SON	DRB	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79516DCQ	ACTIVE	SOT-223	DCQ	6	78	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79516DCQG4	ACTIVE	SOT-223	DCQ	6	78	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79516DCQR	ACTIVE	SOT-223	DCQ	6	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79516DCQRG4	ACTIVE	SOT-223	DCQ	6	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79518DCQ	ACTIVE	SOT-223	DCQ	6	78	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79518DCQG4	ACTIVE	SOT-223	DCQ	6	78	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79518DCQR	ACTIVE	SOT-223	DCQ	6	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79518DCQRG4	ACTIVE	SOT-223	DCQ	6	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79525DCQ	ACTIVE	SOT-223	DCQ	6	78	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79525DCQG4	ACTIVE	SOT-223	DCQ	6	78	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79525DCQR	ACTIVE	SOT-223	DCQ	6	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79525DCQRG4	ACTIVE	SOT-223	DCQ	6	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79530DCQ	ACTIVE	SOT-223	DCQ	6	78	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79530DCQG4	ACTIVE	SOT-223	DCQ	6	78	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79530DCQR	ACTIVE	SOT-223	DCQ	6	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79530DCQRG4	ACTIVE	SOT-223	DCQ	6	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79533DCQ	ACTIVE	SOT-223	DCQ	6	78	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)
TPS79533DCQG4	ACTIVE	SOT-223	DCQ	6	78	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79533DCQR	ACTIVE	SOT-223	DCQ	6	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS79533DCQRG4	ACTIVE	SOT-223	DCQ	6	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR

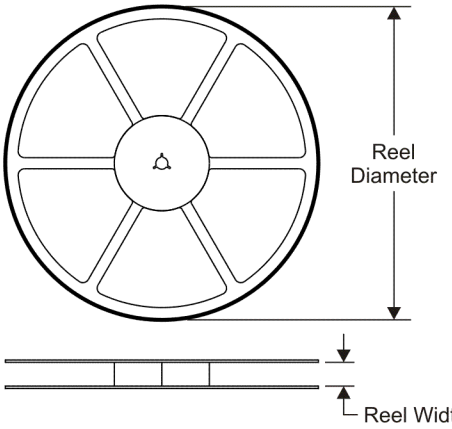
(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.
TBD: The Pb-Free/Green conversion plan has not been defined.
Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.
Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.
Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

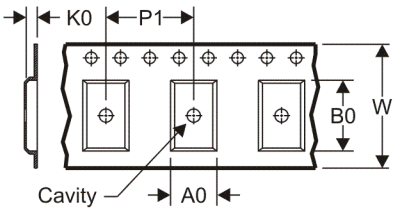
(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

TAPE AND REEL INFORMATION

REEL DIMENSIONS

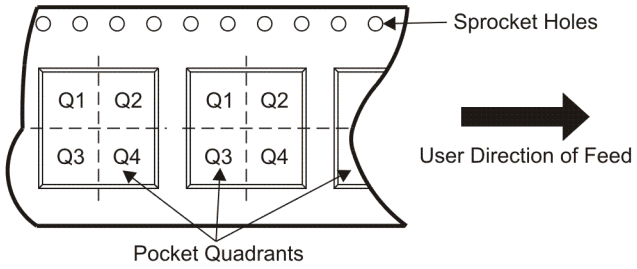


TAPE DIMENSIONS



A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

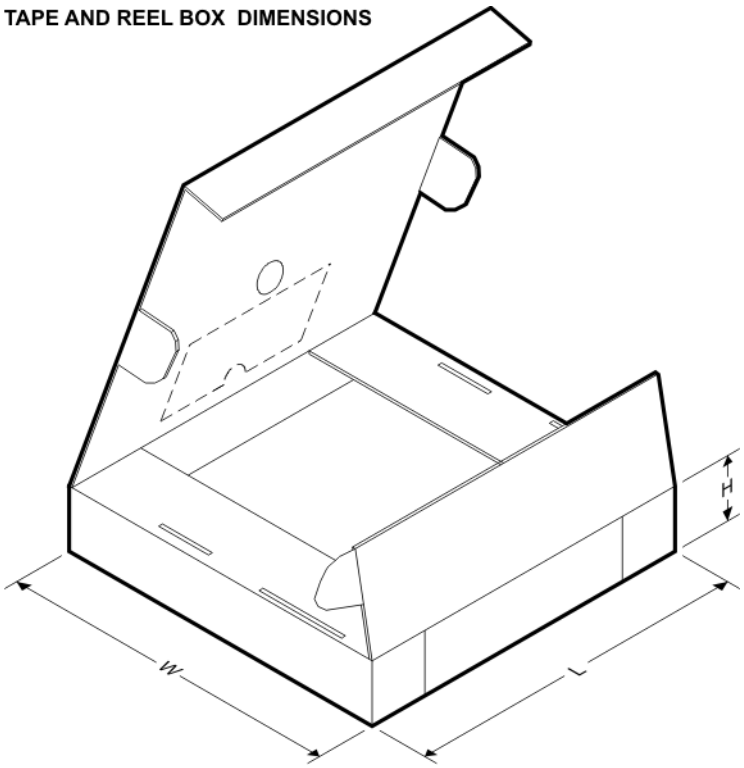
QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS79501DCQR	SOT-223	DCQ	6	2500	330.0	12.4	6.8	7.3	1.88	8.0	12.0	Q3
TPS79501DRBR	SON	DRB	8	3000	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2
TPS79501DRBT	SON	DRB	8	250	180.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2
TPS79516DCQR	SOT-223	DCQ	6	2500	330.0	12.4	6.8	7.3	1.88	8.0	12.0	Q3
TPS79518DCQR	SOT-223	DCQ	6	2500	330.0	12.4	6.8	7.3	1.88	8.0	12.0	Q3
TPS79525DCQR	SOT-223	DCQ	6	2500	330.0	12.4	6.8	7.3	1.88	8.0	12.0	Q3
TPS79530DCQR	SOT-223	DCQ	6	2500	330.0	12.4	6.8	7.3	1.88	8.0	12.0	Q3
TPS79533DCQR	SOT-223	DCQ	6	2500	330.0	12.4	6.8	7.3	1.88	8.0	12.0	Q3

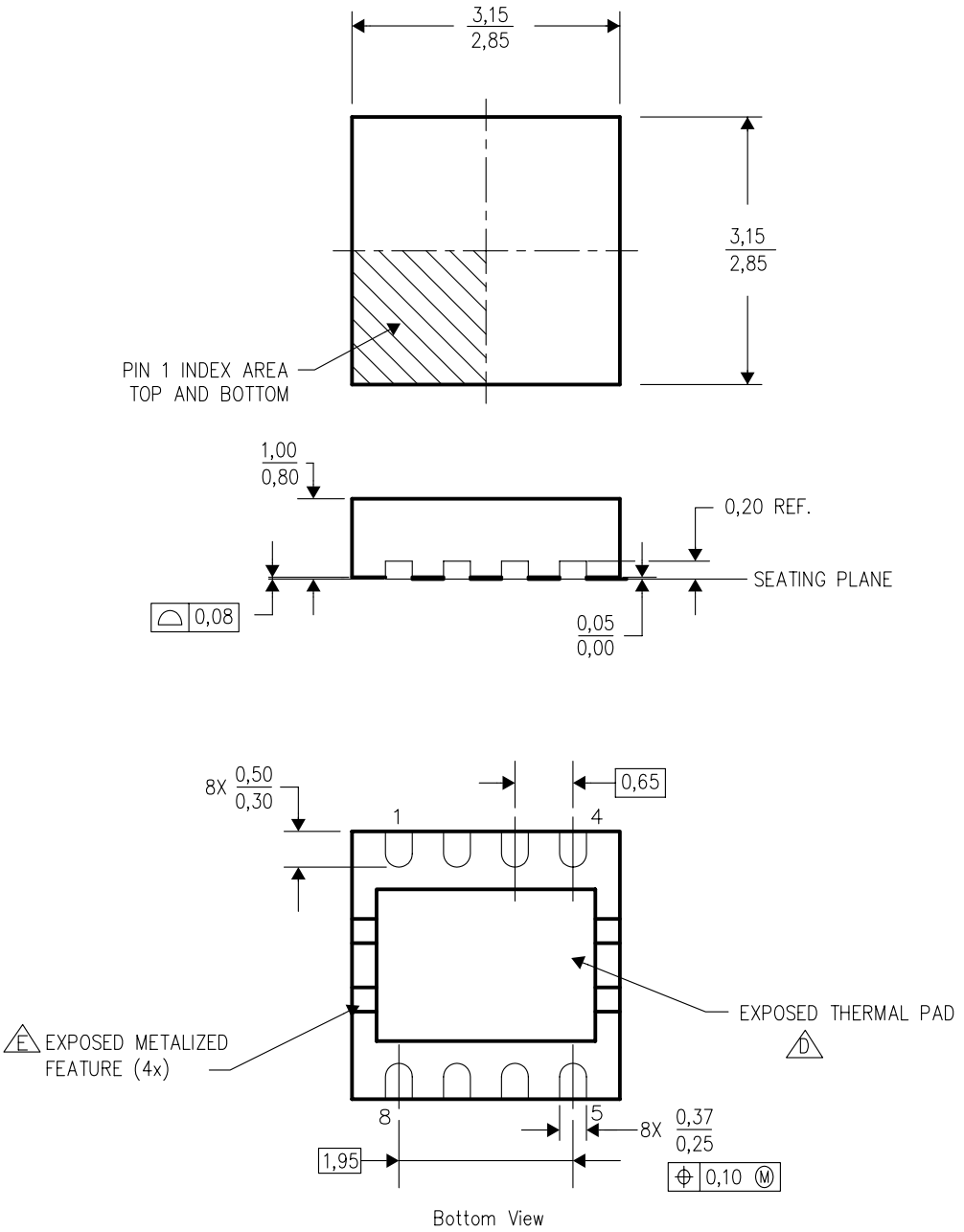
TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS79501DCQR	SOT-223	DCQ	6	2500	358.0	335.0	35.0
TPS79501DRBR	SON	DRB	8	3000	346.0	346.0	29.0
TPS79501DRBT	SON	DRB	8	250	190.5	212.7	31.8
TPS79516DCQR	SOT-223	DCQ	6	2500	358.0	335.0	35.0
TPS79518DCQR	SOT-223	DCQ	6	2500	358.0	335.0	35.0
TPS79525DCQR	SOT-223	DCQ	6	2500	358.0	335.0	35.0
TPS79530DCQR	SOT-223	DCQ	6	2500	358.0	335.0	35.0
TPS79533DCQR	SOT-223	DCQ	6	2500	358.0	335.0	35.0

DRB (S–PDSO–N8) PLASTIC SMALL OUTLINE



4203482/G 11/04

NOTES:
A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M–1994.
B. This drawing is subject to change without notice.
C. Small Outline No–Lead (SON) package configuration.
D. The package thermal pad must be soldered to the board for thermal and mechanical performance. See the Product Data Sheet for details regarding the exposed thermal pad dimensions.
E. Metalized features are supplier options and may not be on the package.

THERMAL PAD MECHANICAL DATA

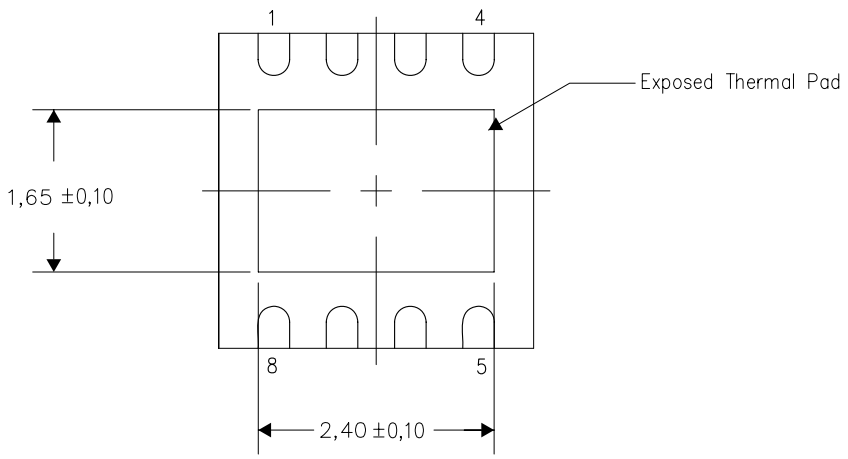
DRB (S–VSON–N8)

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No–Lead (QFN) package and its advantages, refer to Application Report, Quad Flatpack No–Lead Logic Packages, Texas Instruments Literature No. SCBA017. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.

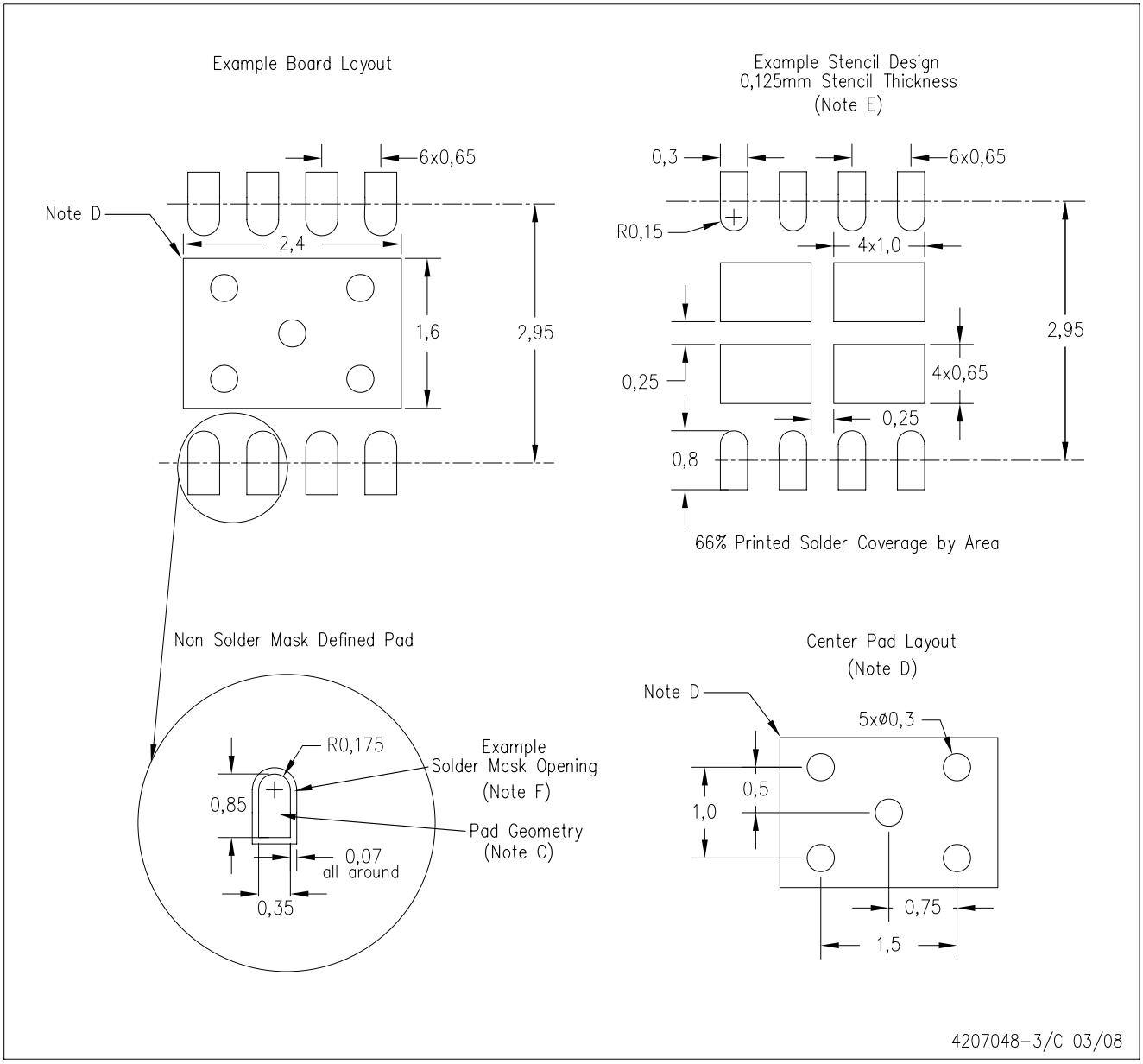


Bottom View

NOTE: All linear dimensions are in millimeters

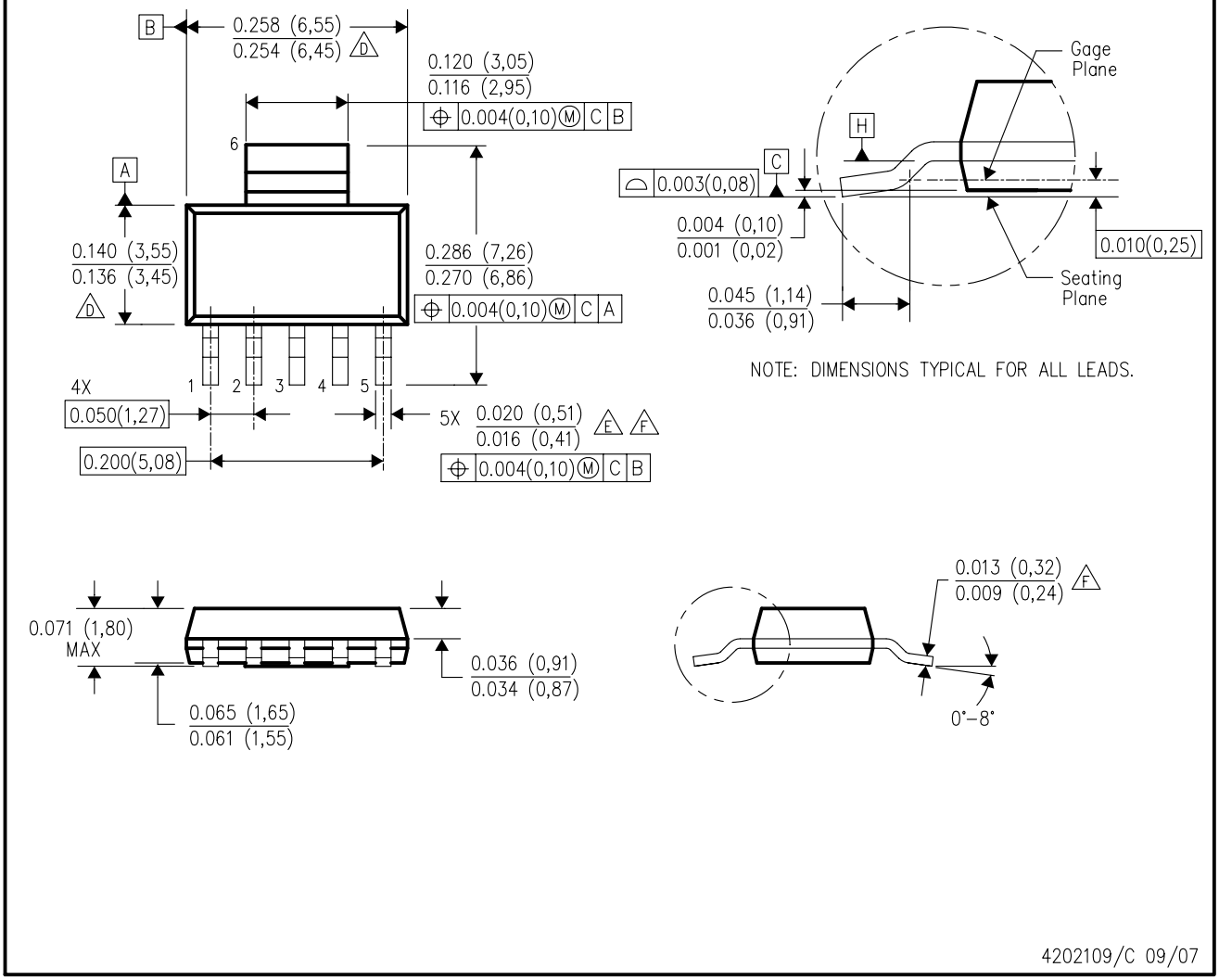
Exposed Thermal Pad Dimensions

DRB (S–VSON–N8)



NOTES:
A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Publication IPC–7351 is recommended for alternate designs.
D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, QFN Packages, Texas Instruments Literature No. SCBA017, SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <http://www.ti.com>.
E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
F. Customers should contact their board fabrication site for solder mask tolerances.

DCQ (R–PDSO–G6) PLASTIC SMALL–OUTLINE



NOTES:
- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- C. Controlling dimension in inches.
- D. Body length and width dimensions are determined at the outermost extremes of the plastic body exclusive of mold flash, tie bar burrs, gate burrs, and interlead flash, but including any mismatch between the top and the bottom of the plastic body.
- E. Lead width dimension does not include dambar protrusion.
- F. Lead width and thickness dimensions apply to solder plated leads.
- G. Interlead flash allow 0.008 inch max.
- H. Gate burr/protrusion max. 0.006 inch.
- I. Datums A and B are to be determined at Datum H.

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